

UEP MIMO-OFDM with Beamforming- Combining for Imperfect Channel Information

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OFDM Workshop, 2007

Outline

- 1 Motivations
 - Motivations for UEP, OFDM, and MIMO
- 2 UEP: Bit-Loading
 - Previous Work
 - Proposed Algorithm
- 3 MIMO-OFDM and Eigen Beamforming
- 4 Simulation Results
 - UEP Adaptive MIMO-OFDM Results
- 5 Conclusions

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Realizing UEP

- ◇ **UEP**: invokes the need for **non-uniform** error protection.
- ◇ **OFDM**: suitable for **adapting individual subcarriers** using different data rates, code rates, and powers
- ◇ **MIMO**: has high **multiplexing** gain and allows for channel layering.
- ◇ **UEP MIMO-OFDM**: devotes an arbitrary number of bits to different classes, eigenbeams, and subcarriers

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Why UEP ?



Equal Error Probability



Unequal Error Probability

- ◇ Source encoders **deliver data of different importance.**
- ◇ Matching the **channel variations** to enhance **performance & capacity.**
- ◇ Different **error sensitivities** of different **communication devices.**

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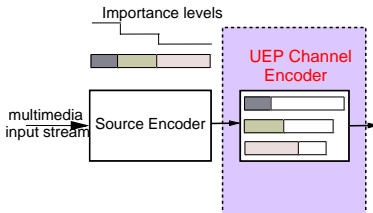
UEP Schemes in MCM

UEP coding layer

Adapt **coding scheme/rate** (i.e., use puncturing or pruning)

UEP physical layer

Adapt **bit/power loading and physical transport, e.g.: MIMO channel**



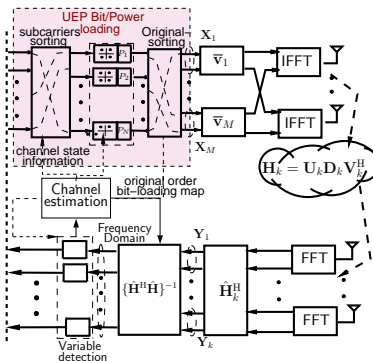
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Bit-Loading Algorithms

Bit-loading solutions:

- **Optimum**: add bits to the locations of minimum incremental power, e.g.: **Hughes-Hartogs** and **Campello**
- **Sub-optimum**: based on Shannon capacity (**Chow et al.**) or probability of error minimization (**Fischer-Huber and Yu-Willson**)

Bit-Loading by Chow (BRMP):

$$b_k = \log_2 \left(1 + \frac{\text{SNR}_k}{\gamma} \right)$$

Quantization Error:

$$\hat{b}_k = \lfloor b_k + 0.5 \rfloor$$

$$\Delta b_k = b_k - \hat{b}_k$$

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MIMO UEP Bit-Loading (BRMP):

$$b_{k,l}^{(j)} = \log_2 \left(1 + \frac{P_{k,l}^{(j)} \cdot \lambda_{k,l}^{(j)}}{\sigma^2 \cdot \gamma^{(j)}} \right)$$

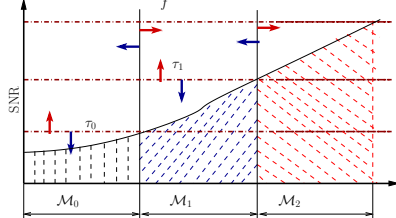
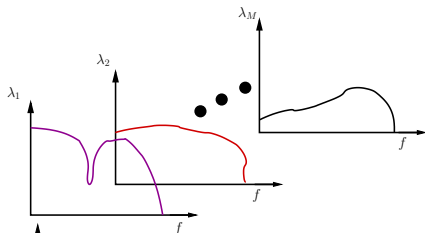
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UEP Bit-Loading and SNR-Sorting Algorithms

- Compute $b_{k,l}^{(j)}$ using $\gamma^{(j)} = \gamma_0 - j \cdot \Delta\gamma$, then adjust $\mathcal{M}^{(j)}$ iteratively until $\sum_{k,l} b_{k,l}^{(j)} = T^{(j)}$ or maximum iteration
- If B_T is not achieved, update γ_0 and recompute. If maximum iterations, add/subtract bits according to $\Delta b_k^{(j)}$
- The power is allocated according to SER. If the target SER is not fulfilled, reduce the total rate

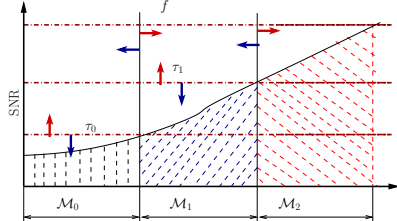
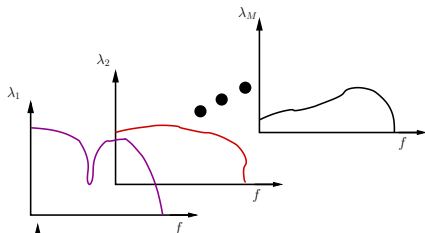


robust-sorting: Class₂(γ_2)  Class₁(γ_1)  Class₀(γ_0) 

intuitive-sorting: Class₀(γ_0)  Class₁(γ_1)  Class₂(γ_2) 

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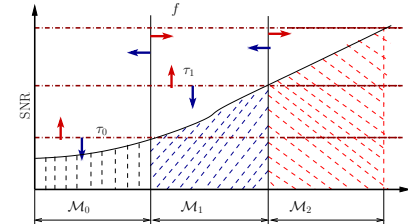
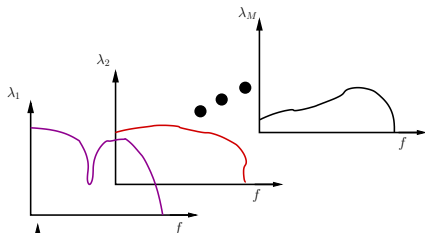


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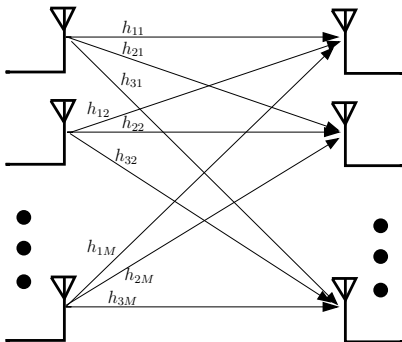
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Potential of MIMO wireless links:

- Substantial improvement in throughput
- Effectively exploit multipath
- Scalability and adaptation

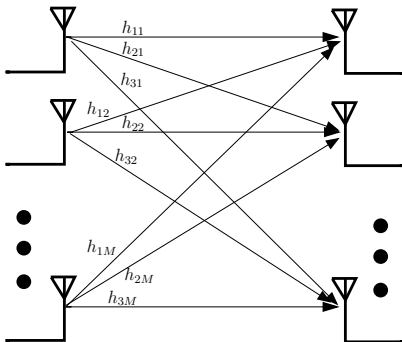


Main Problem!

CSI errors

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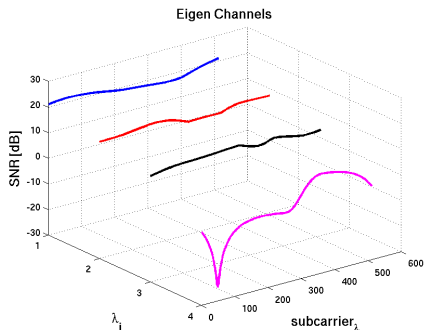


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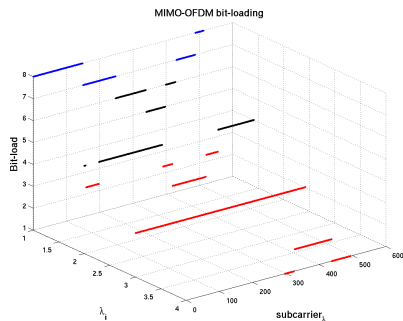
CSI errors

Eigen Channels Representation

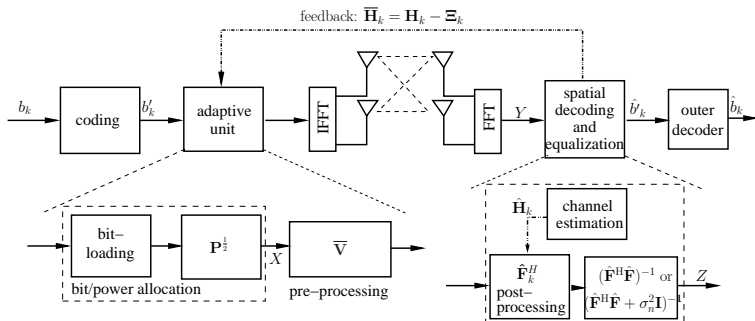
Eigen channels (modes)



Bit-loading for eigen channels



Channel side information feedback



CSI feedback schemes:

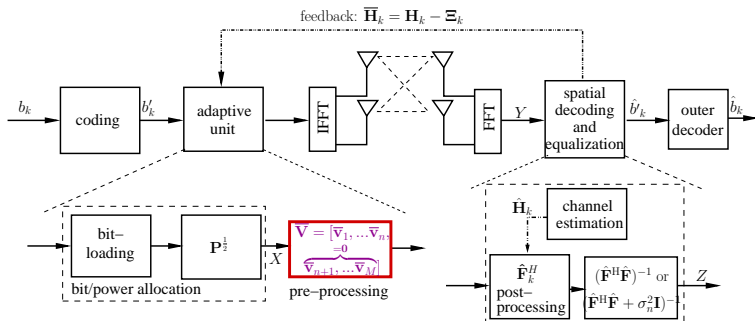
- Channel mean: $\bar{\mathbf{H}} = \mathbf{H} - \mathbf{\Xi}$,
- Channel correlation:

$$\mathbf{R}_{\hat{\mathbf{H}}^H \hat{\mathbf{H}}} = E\{\mathbf{H}^H \mathbf{H}\}$$

CSI uncertainty:

- Channel estimation error
- Quantization error
- errors included by the feedback channel

Channel side information feedback



Beamforming techniques:

- I- full-length beamforming (full-BF), when $n = M$
- II- shorter-length beamforming (n-D BF), due to antenna correlation or CSI errors

System Analysis

CSI error: $\Xi_k = \mathbf{H}_k - \bar{\mathbf{H}}_k$

where $\Xi_k \sim \mathcal{CN}(0, \sigma_{\Xi}^2)$

the received vector:

$$\begin{aligned} \mathbf{Y}_k &= \mathbf{H}_k \bar{\mathbf{V}}_k \mathbf{P}^{1/2} \mathbf{X}_k + n_k \\ &= \underbrace{\mathbf{F}_k}_{\mathbf{U}_k \mathbf{D}_k \mathbf{V}_k^* \bar{\mathbf{V}}_k \mathbf{P}^{1/2}} \mathbf{X}_k + \eta_k, \end{aligned}$$

ZF detection:

$$\mathbf{W} = \{\mathbf{F}^* \mathbf{F}\}^{-1} \mathbf{F}^H$$

MMSE detection:

$$\mathbf{W} = \{\mathbf{F}^* \mathbf{F} + \sigma_N^2 \mathbf{I}\}^{-1} \mathbf{F}^H$$

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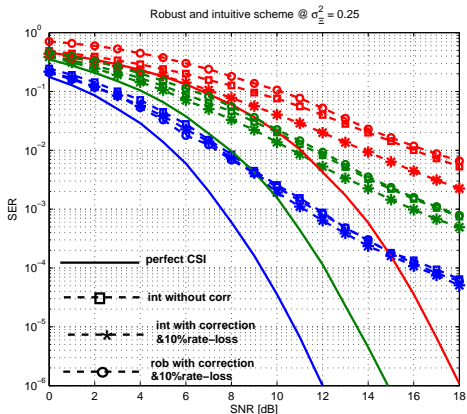
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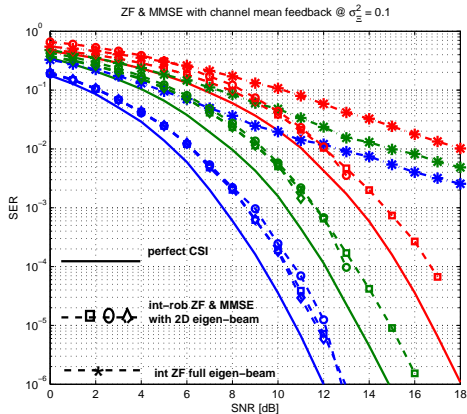
Roubst and intuitive loading using 4D-BF @ $\sigma_{\mathbf{I}}^2 = 0.25$



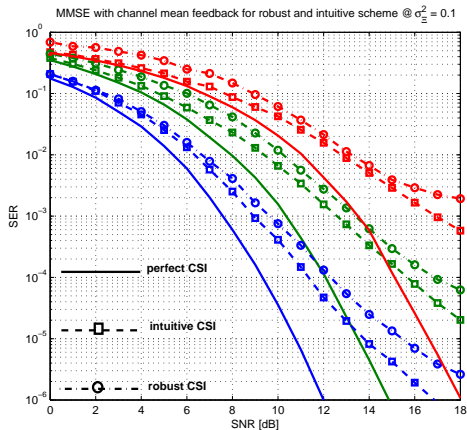
This simulation has been done using:

- ◇ 4 × 4 MIMO-OFDM equivalent to a 512 SISO subcarriers
- ◇ 3 classes, with $\Delta\gamma^{(j)} = 3$ dB, $T^{(j)} = 1024$ bits, and max 8 bits/subcarriers

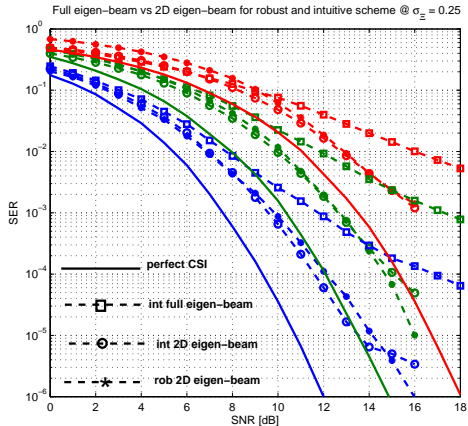
Different equalization using $\sigma_{\underline{\mathbf{H}}}^2 = 0.1$):



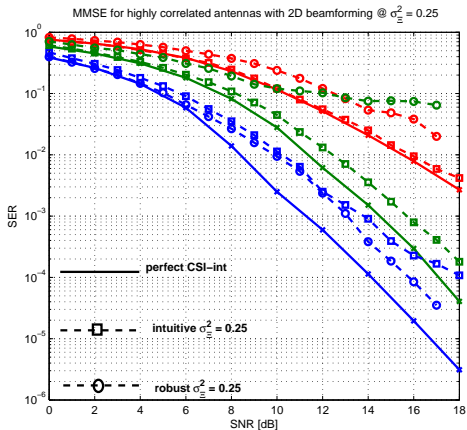
mean CSI using 4D-BF @ $\sigma_{\mathbf{H}}^2 = 0.1$



Roubst and intuitive loading using different beamforming techniques @ $\sigma_{\Xi}^2 = 0.25$)



Highly correlated channels using 2D-BF @ $\sigma_{\Sigma}^2 = 0.25$



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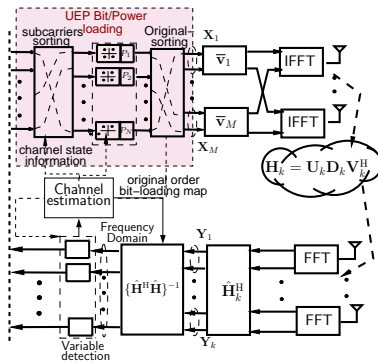
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Ongoing Research:

We are studying the effect of the channel correlation feedback.

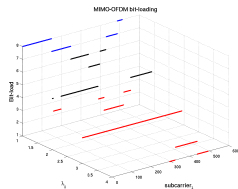
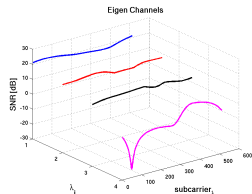


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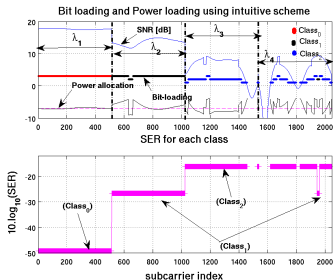


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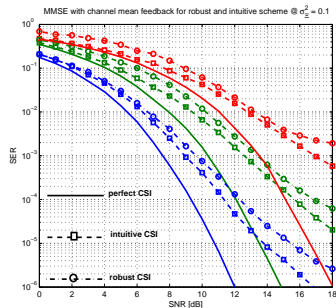
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